

Pb#01(6-3-6): (a) trigonometric Fourier Series $x(t)$.

from C_n plot:

$$\text{at } \omega = 0 \Rightarrow C_0 = 2$$

$$\text{at } \omega = 2 \Rightarrow C_1 = 2$$

$$\text{at } \omega = 3 \Rightarrow C_2 = 1$$

from θ_n plot:

$$\text{at } \omega = 2 \Rightarrow \theta_2 =$$

$$\text{at } \omega = 3 \Rightarrow \theta_3 = -\pi/2$$

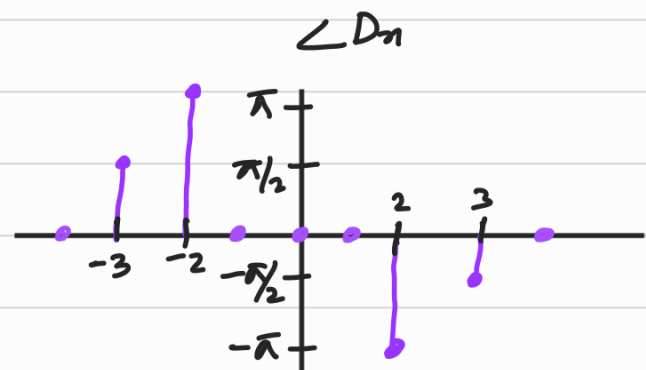
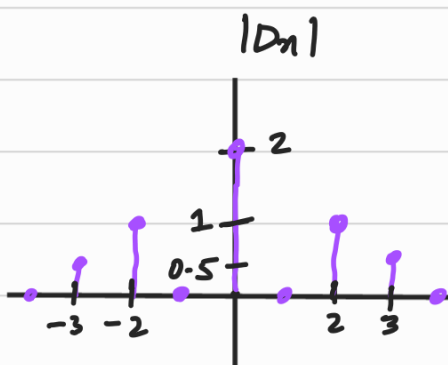
so,

$$x(t) = C_0 + C_1 \cos(2t + \theta_2) + C_2 \cos(3t + \theta_3)$$

$$x(t) = 2 + 2 \cos(2t - \pi) + \cos(3t - \pi/2)$$

$$\boxed{x(t) = 2 - 2 \cos(2t) + \sin(3t)}$$

Pb#01(6-3-6) (b) Sketch Exponential Fourier Spectra.



Pb#01(6-3-6): (c) Exponential Fourier Series for $x(t)$ using sketches in part (b):

$$\Rightarrow x(t) = \sum_{n=-\infty}^{\infty} |D_n| e^{\pm j(n\omega_0 t + \angle D_n)}$$

$$\Rightarrow x(t) = |D_0| + |D_2| e^{-j(2t - \pi)} + |D_{-2}| e^{j(2t - \pi)} +$$

$$|D_3| e^{-j(3t - \pi/2)} + |D_{-3}| e^{j(3t - \pi/2)}$$

$$\Rightarrow x(t) = 2 + e^{j(2t - \pi)} + e^{-j(2t - \pi)} + \frac{1}{2} e^{j(3t - \pi/2)} + \frac{1}{2} e^{-j(3t - \pi/2)}$$

Pb#01 (6.3-6): (d) use $x(t)$ from (c) and apply:

$$\text{Euler's Identity: } e^{-j\theta} + e^{j\theta} = 2 \cos(\theta)$$

$$\Rightarrow x(t) = 2 + 2 \cos(2t - \pi) + \frac{1}{2} (2 \cos(3t - \pi/2))$$

$$\Rightarrow x(t) = 2 + 2 \cos(2t - \pi) + \cos(3t - \pi/2) \quad \checkmark \text{ checked!}$$

Pb#2 (6.3-10): (a) Exponential Fourier Series for $\hat{x}(t) = x(t - T)$

$$\text{since, } x(t) = \sum_{n=-\infty}^{\infty} D_n e^{jn\omega_0 t}$$

$$\text{so, } \hat{x}(t) = x(t - T) = \sum_{n=-\infty}^{\infty} D_n e^{jn\omega_0(t - T)}$$

$$\hat{x}(t) = \sum_{n=-\infty}^{\infty} (D_n e^{-jn\omega_0 T}) e^{jn\omega_0 t} = \sum_{n=-\infty}^{\infty} \hat{D}_n e^{jn\omega_0 t}$$

$$\text{where, } \hat{D}_n = D_n e^{-jn\omega_0 T} \Rightarrow |\hat{D}_n| = |D_n| \text{ and } \angle \hat{D}_n = \angle D_n - jn\omega_0 T$$

done!

Pb#03 (6.4-5): (a) Exponential Fourier Series for periodic signal $x(t)$ (Fig P6.4-5a)

from figure: $\omega_0 = 2\pi$ and,

$$D_n = \int_0^1 e^{-t} e^{-jn\omega_0 t} dt = \int_0^1 e^{(-1-jn\omega_0)t} dt$$

$$D_n = \left. \frac{e^{(-1-jn\omega_0)t}}{-1-jn\omega_0} \right|_0^1 = \frac{e^{-1-jn\omega_0}}{-1-jn\omega_0} - \frac{e^0}{-1-jn\omega_0}$$

$$D_n = \frac{e^{-1-jn\omega_0} - 1}{-1-jn\omega_0} = \frac{e^{-1-jn2\pi} - 1}{-1-jn2\pi}$$

$$D_n = \frac{-(-e^{-1} \cdot e^{-jn2\pi} + 1)}{-(1+jn2\pi)} = \frac{1 - e^{-1}e^{-jn2\pi}}{1+jn2\pi}$$

note: $e^{-jn2\pi} = \cos(2\pi n) - j\sin(2\pi n) = 1$

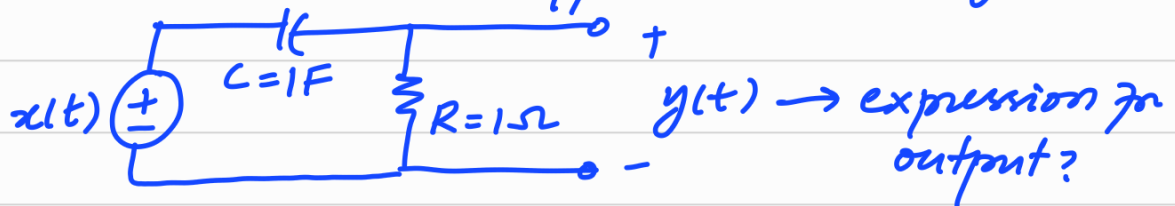
So,

$$D_n = \frac{1 - e^{-1}}{1+jn2\pi}, \text{ And since, } x(t) = \sum_{n=-\infty}^{\infty} D_n e^{jn\omega_0 t}$$

So,

$$x(t) = \sum_{n=-\infty}^{\infty} \left(\frac{1 - e^{-1}}{1+jn2\pi} \right) e^{jn2\pi t}$$

Pb # 03 (6.4-5): (b) $x(t)$ applied to LTIC System



Transfer function of the given RC circuit is

$$H(j\omega) = \frac{1}{1 + \frac{1}{j\omega}} = \frac{j\omega}{j\omega + 1}$$

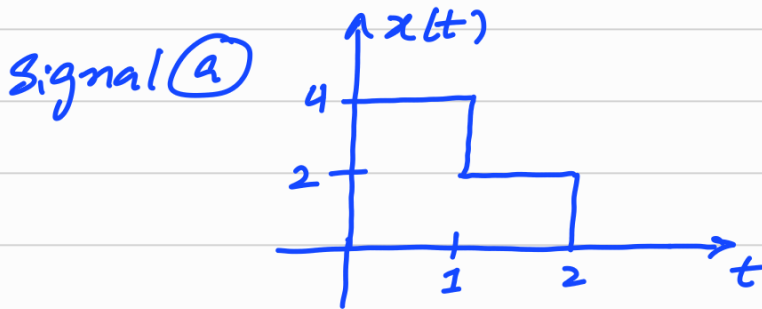
So,

$$\Rightarrow y(t) = \sum_{n=-\infty}^{\infty} D_n H(jn2\pi) e^{jn2\pi t}$$

$$\Rightarrow y(t) = \sum_{n=-\infty}^{\infty} \left(\frac{1 - e^{-1}}{1 + j2\pi n} \right) \left(\frac{j2\pi n}{j2\pi n + 1} \right) e^{j2\pi n t}$$

$$\Rightarrow \boxed{y(t) = \sum_{n=-\infty}^{\infty} \frac{j2\pi n(1 - e^{-1})}{(1 + j2\pi n)^2} e^{j2\pi n t}}$$

Pb#04(7.1-6): Fourier transforms of the signals in Fig P7.1-6

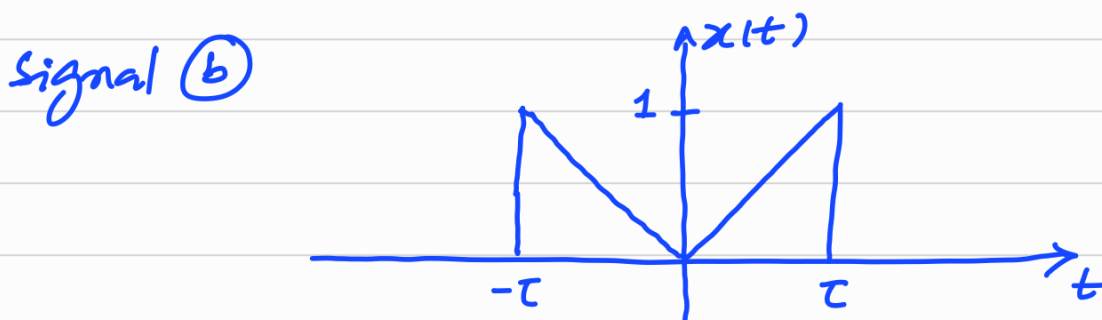


$$X(\omega) = \int_0^1 4 e^{-j\omega t} dt + \int_1^2 2 e^{-j\omega t} dt = \left. \frac{4e^{-j\omega t}}{-j\omega} \right|_0^1 + \left. \frac{2e^{-j\omega t}}{-j\omega} \right|_1^2$$

$$X(\omega) = \frac{4e^{-j\omega}}{-j\omega} - \frac{4e^0}{-j\omega} + \frac{2e^{-j2\omega}}{-j\omega} - \frac{2e^{-j\omega}}{-j\omega}$$

$$X(\omega) = \frac{-4e^{-j\omega} + 4 - 2e^{-j2\omega} + 2e^{-j\omega}}{j\omega}$$

$$\boxed{X(\omega) = \frac{4 - 2e^{-j\omega} - 2e^{-j2\omega}}{j\omega}}$$



Note: $x(t)$ is Even function so,

$$X(\omega) = \frac{2}{T} \int_0^T t \cos(\omega t) dt$$

→ use integration by parts

$$X(\omega) = \frac{2}{T} \left[\frac{t}{\omega} \sin(\omega t) + \frac{1}{\omega^2} \cos(\omega t) \right] \Big|_0^T$$

$$= \frac{2}{T} \left[\frac{T}{\omega} \sin(\omega T) + \frac{1}{\omega^2} \cos(\omega T) - 0 - \frac{1}{\omega^2} \right]$$

$$= \frac{2}{T} \left[\frac{\omega T \sin(\omega T) + \cos(\omega T) - 1}{\omega^2} \right]$$

$$\Rightarrow \boxed{X(\omega) = \frac{2}{T\omega^2} [\cos(\omega T) + \omega T \sin(\omega T) - 1]}$$

Pb#05: Simulink Introduction

LTI System with T.F: $H(s) = \frac{s+2}{s^2+2s+1}$

(a) input $x(t) = 2 \sin(4t)$, $-\infty < t < \infty$
 $y(t) = ?$

Compare $x(t)$ with $x(t) = A \sin(\omega t + \theta)$, gives,
 $A = 2$, $\omega = 4$, $\theta = 0$

and

$$H(s) \rightarrow H(j\omega) = H(j4)$$

$$H(j4) = \frac{j4 + 2}{j^2 16 + 8j + 1} = \frac{j4 + 2}{-16 + j8 + 1} = \frac{j4 + 2}{-15 + j8}$$

so,

$$|H(j4)|^2 = \frac{(2)^2 + (4)^2}{(-15)^2 + (8)^2} = \frac{20}{289} \Rightarrow |H(j4)| = \sqrt{\frac{20}{289}}$$

$$|H(j4)| = \frac{2\sqrt{5}}{17}$$

Also,

$$\angle H(j4) = \tan^{-1}\left(\frac{4}{2}\right) - \tan^{-1}\left(\frac{8}{-15}\right) = \tan^{-1}(2) - \tan^{-1}\left(-\frac{8}{15}\right)$$

$$\angle H(j4) = 63.4349^\circ - (-28.072^\circ + 180^\circ)$$

$$63.4349^\circ - (-28.072^\circ)$$

$$\angle H(j4) = -88.4931^\circ$$

$$= 91.5069$$

so,

$$y(t) = |H(j4)| \cdot 2 \sin(4t + \angle H(j4))$$

so,

$$y(t) = \frac{2\sqrt{5}}{17} \cdot 2 \sin(4t - 88.4931^\circ)$$

$$y(t) = \frac{4\sqrt{5}}{17} \sin(4t - 88.4931^\circ)$$

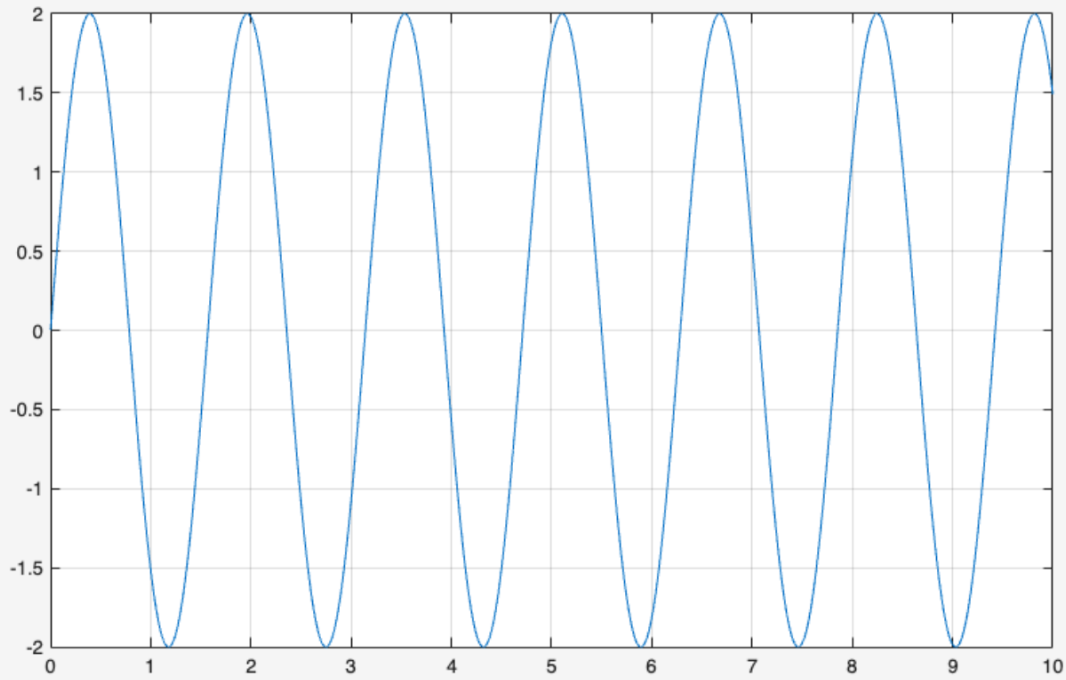
$$y(t) = 0.5261 \sin(4t - 88.4931^\circ)$$

%% problem 5 - part a

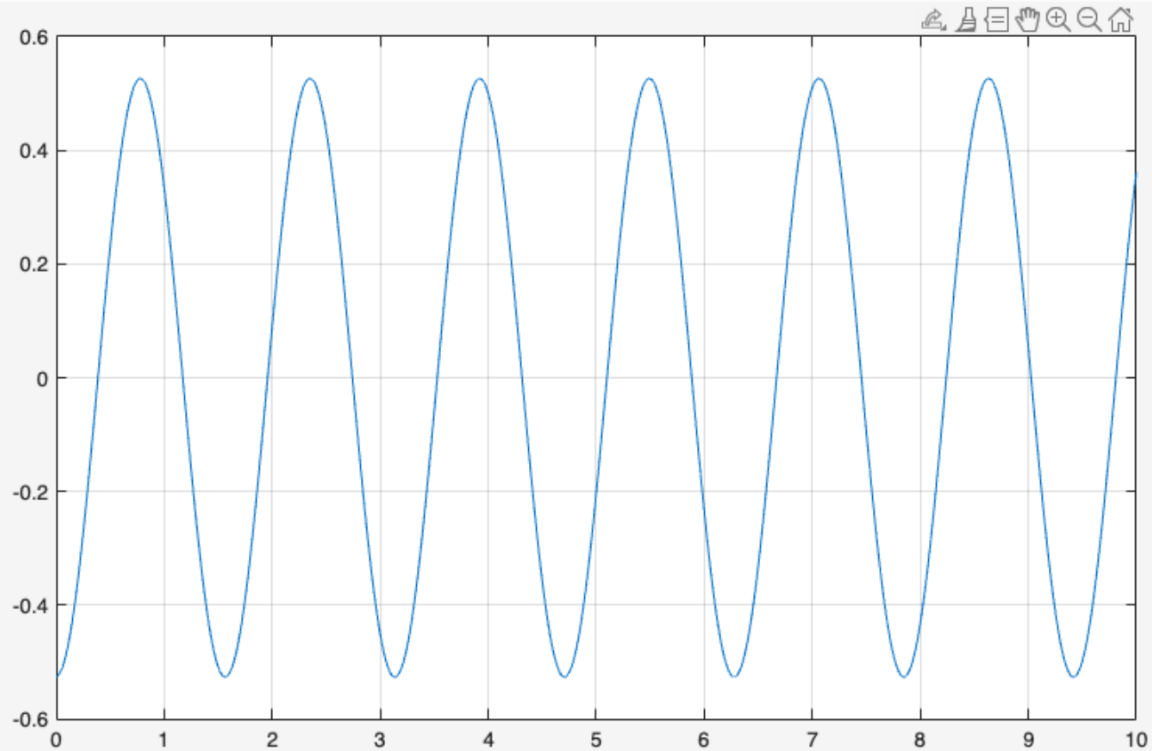
```
t = 0:0.001:10;
x1 = 2*sin(4.*t);
y1_cal = 0.5261*sin(4.*t-88.4931*2*pi/360);
plot(t,x1)
grid on
figure
plot(t,y1_cal)
grid on
```

MATLAB
Code.

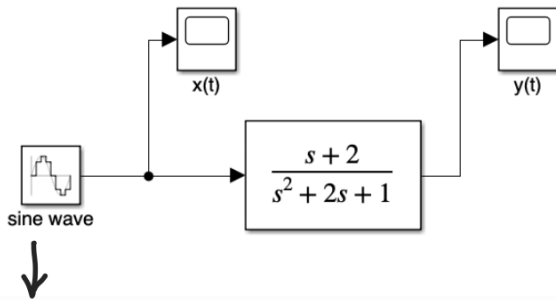
$$x(t) = 2 \sin(4t)$$



calculated $y(t) = 0.5261 \sin(4t - 88.4931^\circ)$



Pb#05 (6) SIMULINK MODEL:



Block Parameters: sine wave

Sine Wave

Output a sine wave:

$O(t) = \text{Amp} * \sin(\text{Freq} * t + \text{Phase}) + \text{Bias}$

Sine type determines the computational technique used. The parameters in the two types are related through:

Samples per period = $2 * \pi / (\text{Frequency} * \text{Sample time})$

Number of offset samples = $\text{Phase} * \text{Samples per period} / (2 * \pi)$

Use the sample-based sine type if numerical problems due to running for large times (e.g. overflow in absolute time) occur.

Parameters

Sine type: Time based

Time (t): Use simulation time

Amplitude: 2

Bias: 0

Frequency (rad/sec): 4

Phase (rad): 0

Sample time: 0.001

☒ Interpret vector parameters as 1-D

OK Cancel Help Apply

→ this is how this sine wave generated block define the signal.

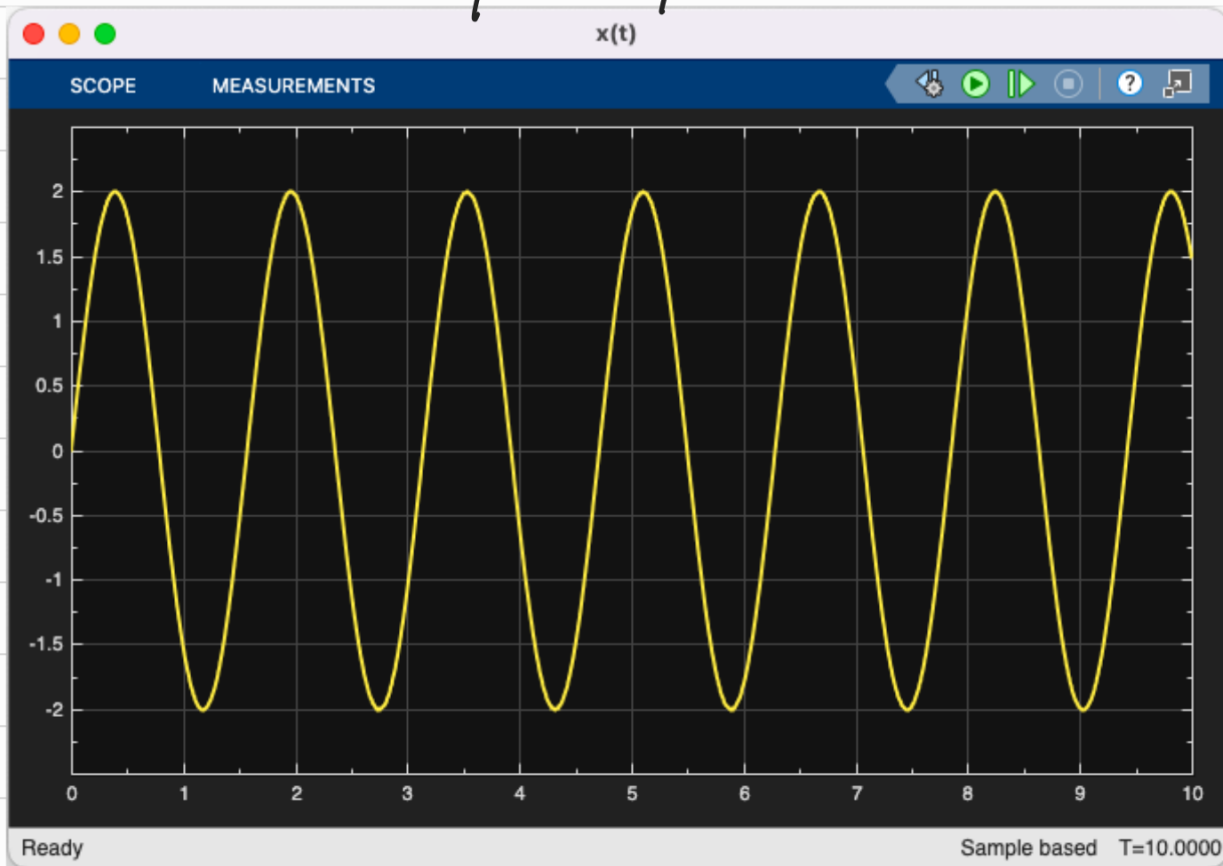
$$\text{So, for } x(t) = 2 \sin(4t)$$

Amplitude = 2

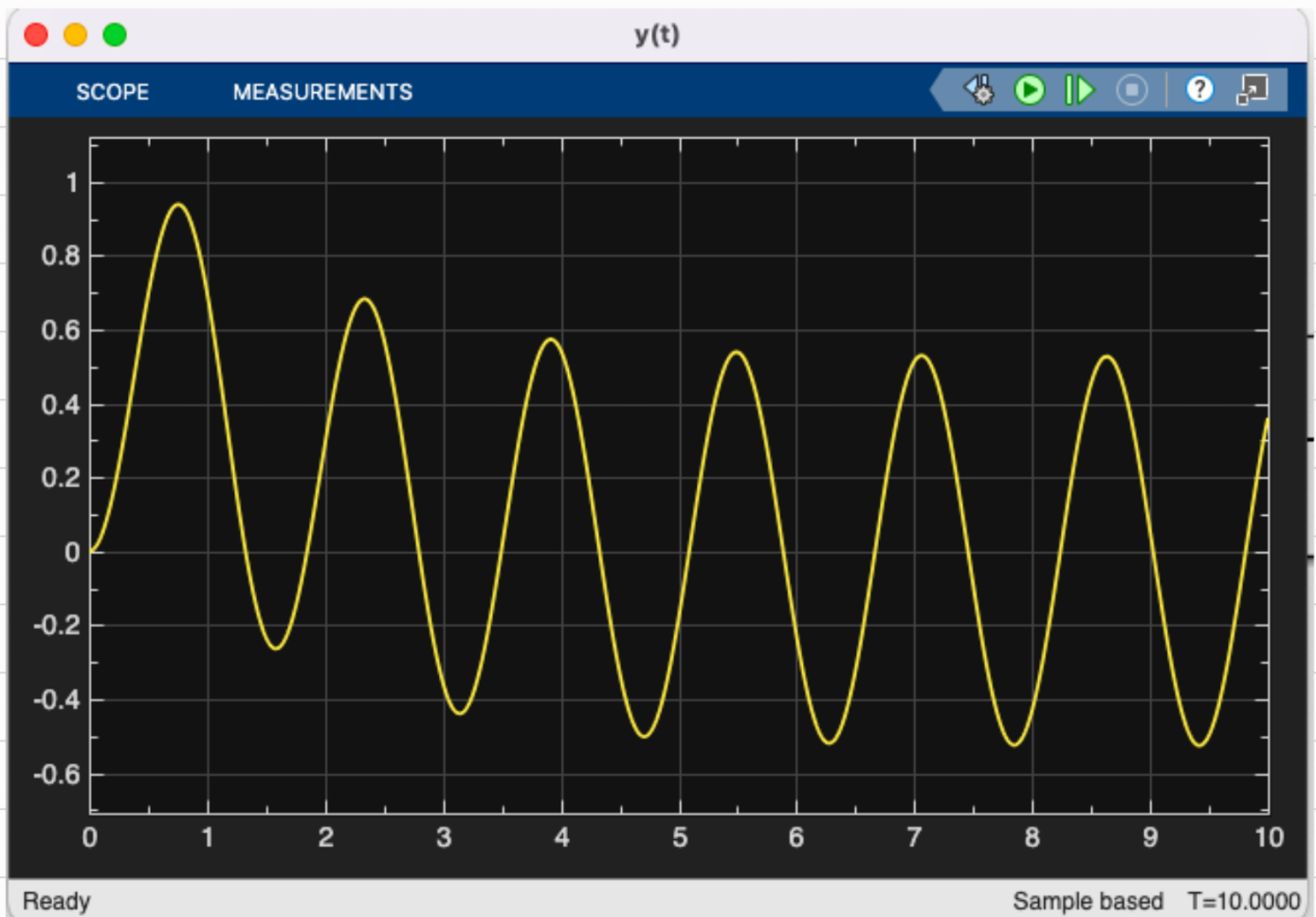
frequency = 4

phase = Bias = 0

Simulink input Scope: $x(t) = 2\sin(4t)$



Simulink output Scope:



Pb#05(c): repeat (a) and (b) for $x(t) = 2\sin(100t)$

So, $A = 2$ same, but $\omega = 100$ now.

So,

$$H(s) \rightarrow H(j\omega) = H(j100)$$

So,

$$H(j100) = \frac{j100 + 2}{10000j^2 + 200j + 1} = \frac{2 + j100}{-9999 + 200j}$$

gives,

$$|H(j100)| = \frac{\sqrt{(2)^2 + (100)^2}}{\sqrt{(-9999)^2 + (200)^2}} = 0.01$$

and

$$\angle H(j100) = \tan^{-1}\left(\frac{100}{2}\right) - \tan^{-1}\left(\frac{200}{-9999}\right)$$

$$= 88.854^\circ - (-1.1458^\circ + 180^\circ)$$

$$88.854^\circ - (-1.1458^\circ)$$

$$\angle H(j100) = -90.0002^\circ$$

$$= 89.9998^\circ$$

So,

$$y(t) = |H(j100)| 2 \sin(100t + \angle H(j100))$$

$$\boxed{y(t) = 0.02 \sin(100t - 90.0002^\circ)}$$

%% problem 5 - part c

`t = 0:0.001:2;` $\rightarrow$ running for less stop time

`x2 = 2*sin(100.*t);`

`y2_cal = 0.02*sin(100.*t-90.0002*2*pi/360);`

`plot(t,x2)`

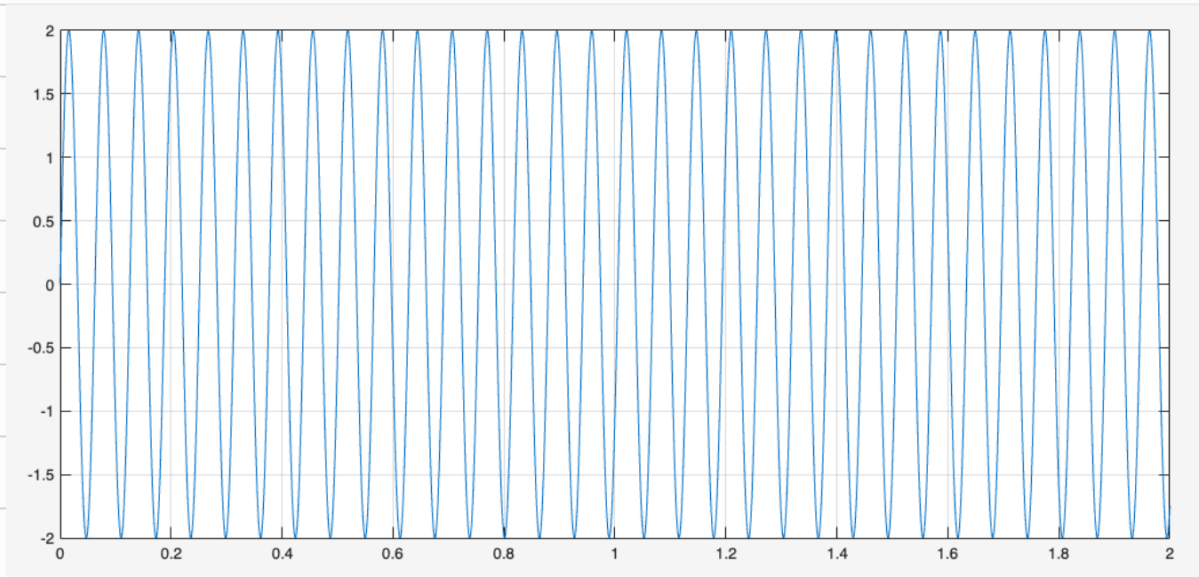
`grid on`

`figure`

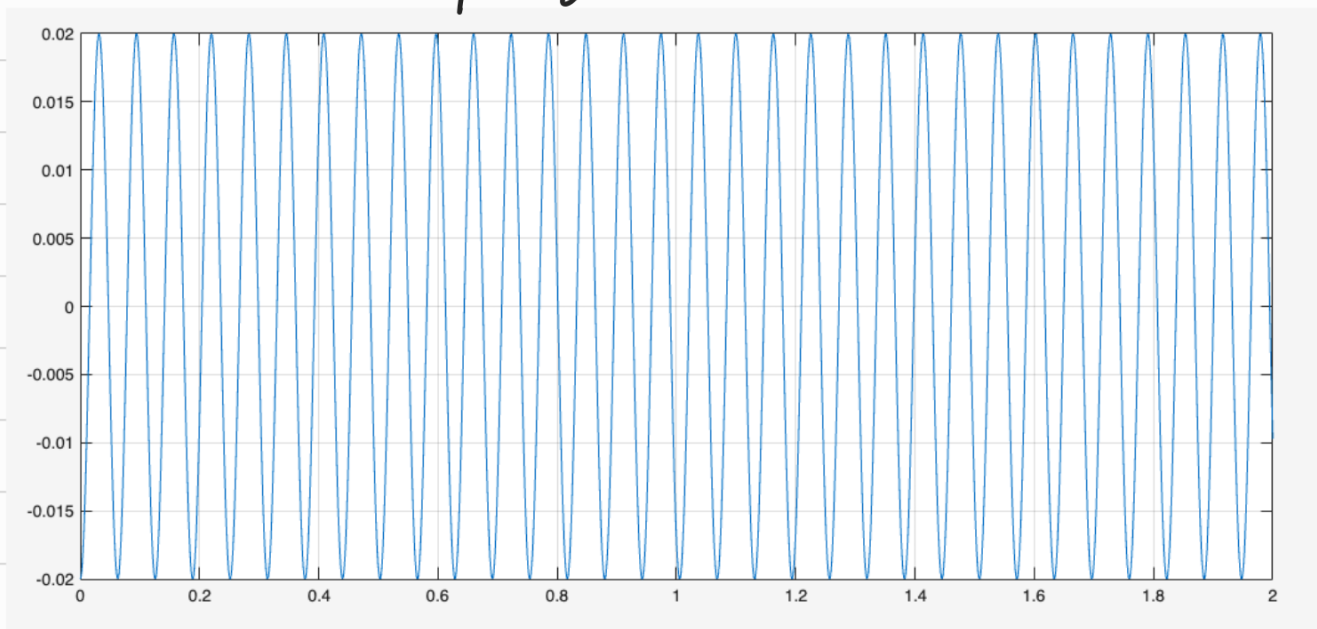
`plot(t,y2_cal)`

`grid on`

input: $x(t) = 2 \sin(100t)$



Calculated output $y(t) = 0.02 \sin(100t - 90.0002^\circ)$



Simulink output Scope result:

